

# Line Impedance and Propagation Coefficient of Narrow Superconducting Coplanar Lines Made of YBaCuO

Jürgen Berntgen, Eberhard Waffenschmidt, Jürgen Musolf, Xiaoyin He, *Member, IEEE*, Michael Heuken, Klaus Heime, *Senior Member, IEEE*, Stefan Hofschien, and Ingo Wolff, *Fellow, IEEE*

**Abstract**—The microwave properties of coplanar waveguides with line widths from 1  $\mu\text{m}$  to 40  $\mu\text{m}$  made of superconducting YBaCuO films with a thickness  $t = 180$  nm on  $\text{LaAlO}_3$  are investigated. The line impedance  $Z_L$  and the normalized propagation coefficient  $\beta/\beta_0$  of these waveguides are measured between 45 MHz and 26.5 GHz at temperatures between 77.4 K and 92 K. The ratio of the line width  $w$  to the distance of the ground layers  $d$  is constant with  $w/d = 0.2$ . Therefore,  $Z_L$  and  $\beta/\beta_0$  are independent of  $w$  for perfectly conducting waveguides. For superconducting waveguides it is found that  $Z_L$  and  $\beta/\beta_0$  differ from the values of perfectly conducting waveguides. They increase for smaller line widths at a constant temperature. At  $w = 1$   $\mu\text{m}$  and  $T = 80$  K,  $Z_L$  and  $\beta/\beta_0$  are nearly twice as high as calculated for perfect conductors. Furthermore,  $Z_L$  and  $\beta/\beta_0$  increase with the temperature. It is shown that these effects are attributed to an increase of the inductance per unit length  $L'$  due to the superconducting material, whereas the capacitance per unit length  $C'$  behaves like  $C'$  of perfectly conducting waveguides. Using these results, the dimensions of the superconducting waveguides, which are necessary to obtain a desired  $Z_L$  at a given line width  $w$ , are calculated.

## I. INTRODUCTION

WHILE THE SURFACE impedance  $Z_S$  is a characteristic quantity for the microwave properties of conducting layers, quantities like the line impedance  $Z_L$ , the propagation coefficient  $\beta$  and the attenuation constant  $\alpha$  play an important role for the characterization of the microwave properties of coplanar waveguides. For normal conductors, many analytical expressions have been published for calculating  $Z_S$ ,  $\alpha$ ,  $\beta$ , and  $Z_L$  [1]–[3]. The surface impedance of superconducting layers shows a different behavior and is normally explained by the two-fluid model and the London equations. For this reason the microwave properties of superconducting coplanar waveguides deviate from those of normal conducting waveguides. No or only a few details are known about the calculation of superconducting coplanar waveguides. This is a severe disadvantage for the design of coplanar transmission lines made of high  $T_C$  superconductors, because unsuitable geometrical dimensions lead to incorrect line impedances  $Z_L$

and propagation coefficients  $\beta$ . This leads to otherwise to avoidable losses because of mismatch or undesirable frequency characteristics of filters and resonators. An exact knowledge of these coplanar line parameters is indispensable for the optimization of superconducting microwave circuit elements, which are to be used in low-loss hybrid superconducting MIC's.

A number of investigations of the microwave properties of YBaCuO thin films has been published to date [4]–[8]. Usually, in most of these papers only the surface impedance  $Z_S$  and the magnetic penetration depth  $\lambda_L$  are determined in order to show the quality of the YBaCuO films. However, until today it is hardly possible to manufacture coplanar waveguides of YBaCuO with exactly defined properties based on this knowledge. There are some publications dealing with the determination of the attenuation constant  $\alpha$  of coplanar waveguides made of YBaCuO [9]–[11]. But in order to give a complete description of the microwave properties the knowledge of  $\beta$  and  $Z_L$  is necessary. Therefore, we focus on the determination of the propagation coefficient  $\beta$  and the line impedance  $Z_L$  of narrow superconducting waveguides. In addition to the effective dielectric constant  $\epsilon_{r\text{eff}}$  we introduce the effective permeability  $\mu_{r\text{eff}}$  in order to give a complete description of the microwave behavior. We have determined  $\epsilon_{r\text{eff}}$  and  $\mu_{r\text{eff}}$  of narrow superconducting coplanar waveguides with line widths above 1  $\mu\text{m}$  from the measured values  $\beta$  and  $Z_L$ . With the results of these investigations it is possible to give an exact prediction of the propagation coefficient and of the line impedance for arbitrary line widths.

## II. THEORETICAL CONSIDERATIONS

The line impedance  $Z_L$  and the propagation coefficient  $\beta$  of a transmission line are dependent on its inductance per unit length  $L'$  and its capacitance per unit length  $C'$ . However, the behavior of  $L'$  and  $C'$  of superconducting structures differs from normally conducting structures. In this chapter, the behavior of superconducting coplanar waveguides is compared to that of perfect conductors and normal conductors.

Using coplanar waveguides made of perfect conductors (infinite conductivity, infinitesimal layer thickness  $t$ ) on a dielectric material with a relative dielectric constant  $\epsilon_r = 1$ , these two quantities only depend on the line width  $w$  and the

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J. Berntgen, E. Waffenschmidt, J. Musolf, X. He, M. Heuken, and K. Heime are with the Institut für Halbleitertechnik, Lehrstuhl I, RWTH Aachen, D 52056 Aachen, Germany

S. Hofschien and I. Wolff are with the Fachgebiet Allgemeine und Theoretische Elektrotechnik, Gerhard-Mercator-Universität GH Duisburg, 47048 Duisburg, Germany

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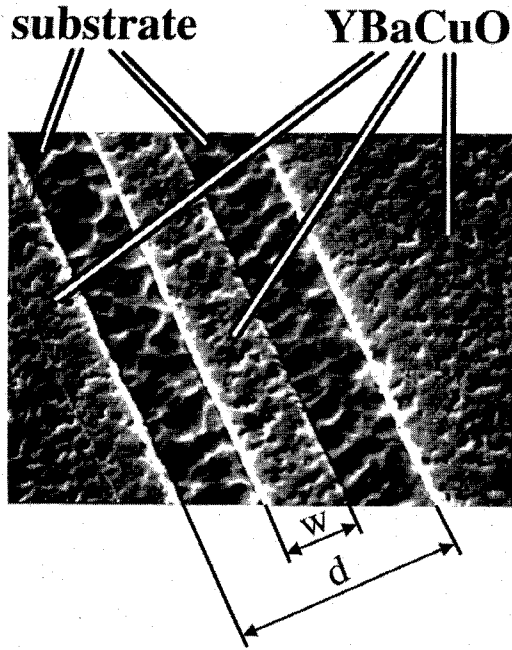


Fig. 1. SEM-photograph of a YBaCuO coplanar waveguide structure.  $w = 1.3 \mu\text{m}$ ,  $d = 4.7 \mu\text{m}$ .

distance  $d$  between the ground layers (compare Fig. 1). The line impedance  $Z_{L\text{geo}}$  and the propagation coefficient  $\beta_0$  of perfect waveguides on a dielectric substrate with  $\epsilon_r = 1$  can be described by the following two equations [3]

$$Z_{L\text{geo}} = 30 \Omega \cdot \pi \cdot \frac{K(k')}{K(k)}, \quad (1)$$

$$\beta_0 = \frac{2\pi f}{c_0}, \quad c_0 = \text{speed of light} \quad (2)$$

with  $k = \frac{w}{d}$ ,  $k' = \sqrt{1 - k^2}$  and  $K(k)$  the complete elliptic first order integral with the module  $k$  or the complementary module  $k'$ , respectively.

Experimentally, the line impedance and the propagation coefficient of normally conducting waveguides are different from (1) and (2). The reason is that  $L'$  and  $C'$  are influenced in a different way by the material properties. In order to describe the influence of the matter on the capacity per unit length  $C'$ , we use the effective dielectric constant  $\epsilon_{r\text{eff}}$ . Additionally, we introduce the effective permeability  $\mu_{r\text{eff}}$ , which gives a description of the inductance per unit length  $L'$ . These two values are given by

$$\epsilon_{r\text{eff}} = \frac{C'}{C'_{\text{geo}}}, \quad (3)$$

$$\mu_{r\text{eff}} = \frac{L'}{L'_{\text{geo}}}. \quad (4)$$

$C'_{\text{geo}}$  and  $L'_{\text{geo}}$  are the capacitance and the inductance per unit length, respectively, of a perfect conducting transmission line using a dielectric substrate with  $\epsilon_r = 1$ ,  $t \rightarrow 0$ .

The definition of  $\epsilon_{r\text{eff}}$  in this work is different from that often used in literature [10], where  $\epsilon_r$  is employed to

describe only the phase velocity or the propagation coefficient. However, our definition of  $\epsilon_{r\text{eff}}$  and  $\mu_{r\text{eff}}$  can be used to describe  $\beta$  and  $Z_L$ . Furthermore, it enables us to separate the capacitive and the inductive influences of the matter on the line properties.

In general, the line impedance  $Z_L$  and the propagation coefficient  $\beta$  of transmission lines made of arbitrary conductors on arbitrary substrates can be determined from  $\epsilon_{r\text{eff}}$  and  $\mu_{r\text{eff}}$  with

$$Z_L = \sqrt{\frac{\mu_{r\text{eff}}}{\epsilon_{r\text{eff}}}} \cdot Z_{L\text{geo}}, \quad (5)$$

$$\beta = \sqrt{\mu_{r\text{eff}} \cdot \epsilon_{r\text{eff}}} \cdot \beta_0. \quad (6)$$

These two equations show that the inductive part of the waveguides has a different effect on  $Z_L$  and  $\beta$  than the capacitive part. Increasing  $\mu_{r\text{eff}}$  increases  $\beta$  as well as  $Z_L$ , but a variation of  $\epsilon_{r\text{eff}}$  influences the two line parameters in an opposite way. It is important to know that a complete determination of  $Z_L$  and  $\beta$  can only be performed with an exact knowledge of  $\epsilon_{r\text{eff}}$  and  $\mu_{r\text{eff}}$ . For that purpose, we concentrated on an analysis of  $\epsilon_{r\text{eff}}$  and  $\mu_{r\text{eff}}$ . In the following chapter we present equations for the theoretical calculation of  $\epsilon_{r\text{eff}}$  and  $\mu_{r\text{eff}}$ , when normal or superconducting coplanar waveguides are used on arbitrary dielectric substrates to underline their different behavior.

A dielectric substrate material with the relative dielectric constant  $\epsilon_r$  has a great influence on  $C'$ . For a coplanar waveguide with a perfectly conducting material on a substrate with  $\epsilon_r$ ,  $\epsilon_{r\text{eff}}$  is given by [3]

$$\epsilon_{r\text{eff}} = \frac{\epsilon_r + 1}{2}. \quad (7)$$

The relative dielectric constant  $\epsilon_r$  of  $\text{LaAlO}_3$  depends on the manufacturing method of the substrate. According to the data sheet of the  $\text{LaAlO}_3$  substrates used in this work [12],  $\epsilon_r$  varies between 20 and 25. The exact value for  $\epsilon_r$  of the batch used in this work is unknown. The properties of  $\text{LaAlO}_3$  have been investigated in detail by Konopka in [13]. They determined a relative dielectric constant  $\epsilon_r = 23.8$  at 77 K, which will be the reference value in this work. Therefore  $\epsilon_{r\text{eff}}$  for perfect conductors on  $\text{LaAlO}_3$  substrate is given by  $\epsilon_{r\text{eff}} = 12.4$ .

Normal conducting coplanar waveguides with low losses and a finite layer thickness behave similar to perfect conductors. Therefore, it is possible to describe their microwave properties by using  $\epsilon_{r\text{eff}}$  only. The coefficient  $\mu_{r\text{eff}}$  is closed to one.

No analytic expression for  $C'$  of superconducting coplanar lines has been published so far, therefore the characteristics of  $\epsilon_{r\text{eff}}$  must be examined in more detail.

Contrary to normal conductors, waveguides made of superconducting materials show a significant contribution to  $L'$  [9], [14], [15]. Therefore, the effective permeability  $\mu_{r\text{eff}}$  cannot be neglected. In [14] an equation is quoted for calculating the inductance per unit length  $L'$  of a superconducting coplanar waveguide which uses the geometrical dimensions and the magnetic penetration depth  $\lambda_L$ . Using this equation the

effective permeability  $\mu_{r\text{eff}}$  is given by

$$\mu_{r\text{eff}} = 1 + \frac{\lambda_L(T)}{K(k')} \cdot \frac{C}{A \cdot D} \cdot \left\{ \frac{1,7}{\sinh\left(\frac{t}{2\lambda_L(T)}\right)} + \frac{0,4}{\sqrt{E}} \right\} \quad (8)$$

with

$$\begin{aligned} A &= -\frac{t}{\pi} + \frac{1}{2} \cdot \sqrt{\left(\frac{2 \cdot t}{\pi}\right)^2 + w^2}, & B &= \frac{w^2}{4 \cdot A} \\ C &= B - \frac{t}{\pi} + \sqrt{\left(\frac{t}{\pi}\right)^2 + \frac{(d-w)^2}{4}}, & D &= \frac{2 \cdot t}{\pi} + C \\ E &= \left[ \left(\frac{B}{A}\right)^2 - 1 \right] \cdot \left[ 1 - \left(\frac{B}{D}\right)^2 \right]. \end{aligned}$$

The coefficient  $\mu_{r\text{eff}}$  shows a temperature dependence, which results from the temperature dependence of  $\lambda_L(T)$ . In this publication  $\lambda_L(T)$  was determined from cavity resonator measurements performed on the unpatterned sample. The temperature dependence of  $\lambda_L(T)$  agreed well to the London theory and the thermodynamic model by Gorter and Casimir [16]

$$\lambda_L(T) = \lambda_L(0) \cdot \frac{1}{\sqrt{1 - \left(\frac{T}{T_c}\right)^4}}. \quad (9)$$

Due to the frequency independence of  $\lambda_L(T)$  it is expected that  $\mu_{r\text{eff}}$  is also independent of the frequency.

According to (8),  $\mu_{r\text{eff}}$  is dependent on the magnetic penetration depth  $\lambda_L(T)$ . If the geometric dimensions of the waveguide are known, it is possible to determine  $\lambda_L$  from the measured value for  $\mu_{r\text{eff}}$ . Equation (8) can be transformed to  $\lambda_L$  using the approximation  $\sinh(x) \approx x$  with  $x = t/(2\lambda_L) < 1$ . For  $x < 0.4$  the deviation from the exact value becomes less than 1%. This deviation can be neglected for samples used in this work, which have a film thickness of about 200 nm. Hence it follows for  $\lambda_L$

$$\lambda_L = -\frac{t}{17 \cdot \sqrt{E}} \cdot \left[ \frac{t^2}{17^2 \cdot E} + (\mu_{r\text{eff}} - 1) \cdot t \cdot K(k') \cdot \frac{A \cdot D}{3,4 \cdot C} \right]^{\frac{1}{2}} \quad (10)$$

with  $A$ ,  $B$ ,  $C$ ,  $D$ , and  $E$  according to (8).

With this expression the material constant  $\lambda_L(0)$  of superconducting coplanar waveguides can be determined according to (9).

### III. EXPERIMENTAL PROCEDURE

#### A. Film Deposition and Device Technology

Epitaxial *c*-axis oriented YBaCuO thin films were deposited by metal organic chemical vapor deposition (MOCVD) on 10 mm × 10 mm LaAlO<sub>3</sub> (100) substrates with a substrate thickness of  $h = 0.5$  mm. The YBaCuO thin films were grown at temperatures between 1023 K and 1073 K in a commercial AIXTRON horizontal reactor with cold walls [17]. Samples

with a critical current density  $j_C$  (77 K, 0 T)  $> 2 \cdot 10^6$  A/cm<sup>2</sup> and a microwave surface resistance  $R_s$  (10 GHz, 77 K)  $\approx 0.2$  mΩ could be obtained [4], [18]. The thickness of the superconducting films used in this work is about  $t = 180$  nm.

The films were patterned into coplanar waveguide structures using a conventional photolithographic process and Ar ion-etching. Using this process superconducting transmission lines with line widths from 40 μm down to  $w = 1$  μm (see Fig. 1) were manufactured. The film showed no significant degradation, because the critical temperature  $T_C$  deteriorated only less than 1 K.

In order to obtain a contact resistance of less than  $10^{-6}$  Ωcm<sup>2</sup> the following process was used: The contacts made of palladium (thickness: 25 nm) and gold (thickness: 225 nm) were evaporated onto the YBaCuO and patterned by lift-off. The contacts were annealed in an oxygen atmosphere at 650°C for one hour and then slowly cooled down.

The coplanar waveguide structures are transmission lines (geometrical length = 1.7 mm) and  $\lambda/2$ -resonators (geometrical length = 7.4 mm). The ratio between the line width  $w$  and the distance from the ground layers  $d$  was chosen to  $k = w/d = 0.2$ . For coplanar waveguides consisting of perfect conductors on LaAlO<sub>3</sub> ( $\epsilon_r = 23.8$ ) this ratio would lead to a line impedance  $Z_L = 50$  Ω.

#### B. Microwave Measurements

The coplanar microwave structures were characterized by measuring the scattering parameters between 45 MHz and 26.5 GHz by a network analyzer HP8510. The system was cooled by liquid nitrogen and allowed measurements in a temperature range from 77.4 K to 300 K with a temperature accuracy of  $\pm 0.1$  K. For contacting the samples, two commercial microwave probes were used, which allowed on-wafer measurements. The line impedance of the test system is  $Z_{L0} = 50$  Ω.

The two unknown line parameters  $\beta$  and  $Z_L$  or  $\epsilon_{r\text{eff}}$  and  $\mu_{r\text{eff}}$ , respectively, were determined from the measured scattering parameters.

Mismatched transmission lines ( $Z_L \neq Z_{L0}$ ) show a reflection parameter  $S_{11}$ , the amount of which shows distinct maxima and minima as a function of frequency. If the line is terminated by  $Z_{L0}$ , the maxima  $|S_{11\text{max}}|$  occur at the phase  $\varphi = \arg(S_{11}) = 0$  for line impedances bigger than  $Z_{L0}$ , or at the phase  $\varphi = \arg(S_{11}) = \pi$  for  $Z_L < Z_{L0}$ . At the first maximum of  $|S_{11\text{max}}|$  the transmission line has a length of exactly  $1 = \lambda/4$ , and the line impedance of the mismatched transmission line can be determined by a  $\lambda/4$ -transformation as follows

$$Z_L = Z_{L0} \cdot \sqrt{\frac{1 + S_{11\text{max}}}{1 - S_{11\text{max}}}}. \quad (11)$$

The propagation coefficient  $\beta$  is determined from the resonant frequencies  $f_1 \cdots f_n$  of the resonator with  $n$  giving the wave number of the resonance mode. It must be considered that the capacitive coupling of the resonator causes a virtual prolongation  $l_{\text{virt}}$  because of the coupling slots. The prolongation  $l_{\text{virt}}$  is determined by a separate measurement of various coupling

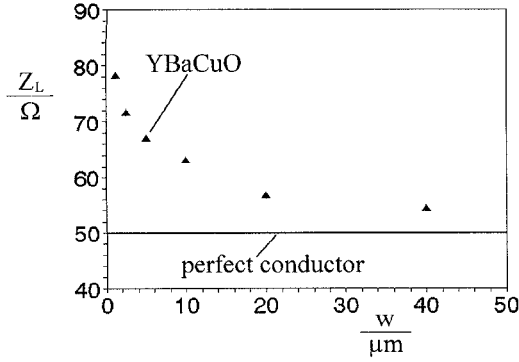


Fig. 2. Line impedance  $Z_L$  of a YBaCuO coplanar line on LaAlO<sub>3</sub> substrate in dependence on the line width  $w$  with  $w/d = 0.2$ ,  $t = 180$  nm,  $T = 80$  K,  $\lambda_L(80 \text{ K}) = 400$  nm.

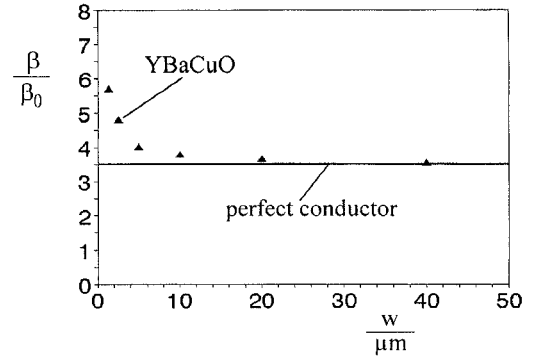


Fig. 3. Normalized propagation coefficient  $\beta/\beta_0$  of a YBaCuO coplanar line on LaAlO<sub>3</sub> substrate in dependence on the line width  $w$  with  $w/d = 0.2$ ,  $t = 180$  nm,  $T = 80$  K,  $\lambda_L(80 \text{ K}) = 400$  nm.

slots existing on the sample. The electrically effective length of a capacitively coupled  $\lambda/2$ -resonator is  $l_{\text{eff}} = l_{\text{geo}} + l_{\text{virt}}$ . So the propagation coefficient is determined by the following equation

$$\beta_n = \frac{n \cdot \pi}{l_{\text{eff}}}. \quad (12)$$

With these two line parameters the effective dielectric constant  $\epsilon_{r \text{ eff}}$  and the effective permeability  $\mu_{r \text{ eff}}$  were determined by rearranging (5) and (6)

$$\epsilon_{r \text{ eff}} = \frac{\beta_n}{\beta_0} \cdot \frac{Z_{L \text{ geo}}}{Z_L}, \quad (13)$$

$$\mu_{r \text{ eff}} = \frac{\beta_n}{\beta_0} \cdot \frac{Z_L}{Z_{L \text{ geo}}}. \quad (14)$$

Additionally  $\epsilon_{r \text{ eff}}$  and  $\mu_{r \text{ eff}}$  were simulated by the two-dimensional finite difference time domain method (2-D-FDTD) described in [19]. This simulation considers the London equations assuming a penetration depth  $\lambda_L(T) = 400$  nm for the superconductor, a relative dielectric constant  $\epsilon_r = 23.8$  for the substrate and a thickness of the layer  $t = 200$  nm.

#### IV. RESULTS AND DISCUSSION

Experimental results for the line impedance  $Z_L$  and the propagation coefficient  $\beta$  of coplanar YBaCuO structures are shown in Figs. 2 and 3 and compared to values calculated for perfect conductors. Despite a constant ratio  $w/d$ , both  $Z_L$  and  $\beta$  show a significant increase for line widths below  $20 \mu\text{m}$ . This behavior cannot be explained by a deviation from exact dimensions due to a limited resolution of the photolithographic process or due to an isolating part at the line edges. This would only cause a variation of  $Z_L$ , but  $\beta/\beta_0$  would not be affected. According to the definitions of the two line parameters  $Z_L$  (5) and  $\beta$  (6) we conclude that a variation of  $\mu_{r \text{ eff}}$  is more dominant than a variation of  $\epsilon_{r \text{ eff}}$  at narrow line widths.

From these results  $\epsilon_{r \text{ eff}}$  and  $\mu_{r \text{ eff}}$  were calculated using (13) and (14) in order to resolve the influence of  $L'$  and  $C'$  on  $Z_L$  and  $\beta/\beta_0$ . The dependence of  $\epsilon_{r \text{ eff}}$  and  $\mu_{r \text{ eff}}$  on the line width are presented in Figs. 4 and 5.

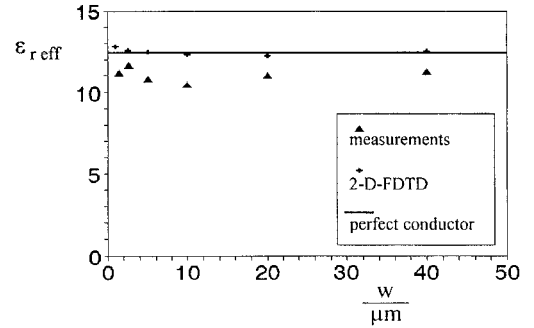


Fig. 4. Effective dielectric constant  $\epsilon_{r \text{ eff}}$  of a YBaCuO coplanar line on LaAlO<sub>3</sub> substrate in dependence on the line width  $w$  with  $w/d = 0.2$ ,  $t = 180$  nm,  $T = 80$  K,  $\lambda_L(80 \text{ K}) = 400$  nm.

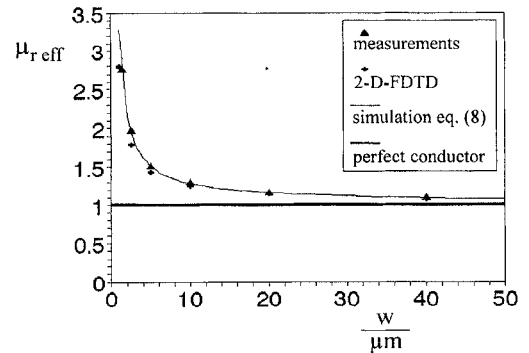


Fig. 5. Effective permeability  $\mu_{r \text{ eff}}$  of a YBaCuO coplanar line on LaAlO<sub>3</sub> substrate in dependence on the line width  $w$  with:  $w/d = 0.2$ ,  $t = 180$  nm,  $T = 80$  K,  $\lambda_L(80 \text{ K}) = 400$  nm.

Fig. 4 shows the measured  $\epsilon_{r \text{ eff}}$  for coplanar structures made of YBaCuO on LaAlO<sub>3</sub> in dependence on the line width  $w$ , compared to the expected  $\epsilon_{r \text{ eff}}$  for a perfect conductor, and a simulation for the YBaCuO sample performed with the 2-D-FDTD method. The 2-D-FDTD method and the calculation for a perfect conductor show nearly identical results.

The measured  $\epsilon_{r \text{ eff}}$  is nearly independent of the line width and is to be regarded as constant with an average value of about 11.4. This average value, which has been determined from several samples, is below the expected value  $\epsilon_{r \text{ eff}} = 12.4$

for a perfect conductor and lower than the simulation for superconductors with the 2-D-FDTD method. This deviation might be traced to the properties of the substrate. According to the data sheet [12] of the  $\text{LaAlO}_3$  substrate, the relative dielectric constant  $\epsilon_r$  of different batches may vary between 20 and 25. This effect is caused by fluctuations of the manufacturing method. Using (7) this leads to  $\epsilon_{r\text{eff}}$  values between 10.5 and 13. The result of the measurements shows that it is important to determine exactly the dielectric properties of the batch of substrates in order to give an exact description of the microwave properties of the superconductor. On the contrary, the effective permeability  $\mu_{r\text{eff}}$  of YBaCuO coplanar transmission lines determined from (14) shows a strong increase to narrow line widths (Fig. 5). This is due to a large variation of the inductance per unit length  $L'$ , which is caused by the inertia of the superconducting electrons. It can be explained by the two-fluid model and the London equations [9], [14]. Microwave structures made of superconducting materials show a different behavior compared to those made of a perfect conductor, because their  $\mu_{r\text{eff}}$  differs widely from the value for perfect materials at line widths below  $w = 15 \mu\text{m}$ . Additionally, in Fig. 5 two different simulations for  $\mu_{r\text{eff}}$  of YBaCuO are shown with  $t = 180 \text{ nm}$  and  $\lambda_L(T) = 400 \text{ nm}$ : the 2-D-FDTD method and the determination using (8). The 2-D-FDTD simulation shows slight deviations from the measured results, but it should be taken into account that only the originally given conductor geometry and the material parameters for a dielectric constant of the substrate  $\epsilon_r = 24.8$ ,  $\lambda_L(0)$  and  $T_C$  have been used. The measured values correspond well to those determined from (8). Therefore, an analytic expression (8) for  $\mu_{r\text{eff}}$  has been found, using only the geometrical dimensions and the magnetic penetration depth  $\lambda_L(T)$  of the superconducting material.

Figs. 4 and 5 illustrate the dependence of the line width on both, the line impedance  $Z_L$  and the propagation coefficient  $\beta$ , only determined by  $\mu_{r\text{eff}}$ , respectively  $L'$ . In addition to the dimensions of the waveguide, the quantities like film thickness  $t$  and the penetration depth  $\lambda_L(T)$  of the superconductor exhibit an effect on  $\mu_{r\text{eff}}$ . Due to the deviation caused by the manufacturing method, both quantities are not exactly reproducible. This correlation causes some problems in manufacturing waveguides with exact  $Z_L$  and  $\beta$ . Small variations of the dimensions ( $w$ ,  $d$ ,  $t$ ,  $\lambda_L$ ) cause a large deviation from the desired  $Z_L$  and  $\beta$  values at narrow line widths below  $w = 5 \mu\text{m}$ .  $Z_L$  and  $\beta$  scatter in a wide range for these structures. This deviation is less significant for practical applications with line widths above  $w = 10 \mu\text{m}$ .

Moreover, the line parameters  $\beta$  and  $Z_L$  are strongly influenced by the temperature. In Fig. 6  $\epsilon_{r\text{eff}}$  and  $\mu_{r\text{eff}}$  are shown as a function of the temperature for a YBaCuO coplanar line with  $T_C = 90.3 \text{ K}$  and  $\lambda_L(0) = 286 \text{ nm}$ . The effective dielectric constant  $\epsilon_{r\text{eff}}$  shows only a very small temperature dependence in the measured temperature range, even near  $T_C$ . This can be explained by the dielectric characteristics of the substrate  $\text{LaAlO}_3$  [13]. The relative dielectric constant  $\epsilon_{r\text{eff}}(T)$  is regarded as a constant in this small temperature range. Contrary,  $\mu_{r\text{eff}}$  increases significantly with increasing the temperature. This behavior can be explained by (8) and

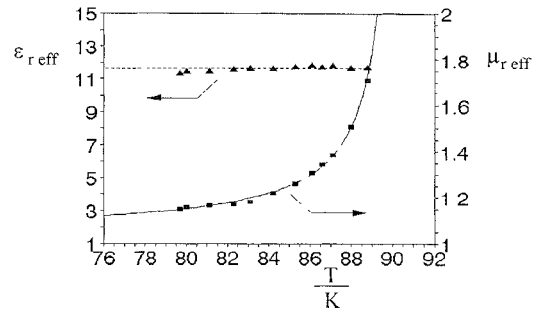


Fig. 6. Measured values of the effective dielectric constant  $\epsilon_{r\text{eff}}$  and effective permeability  $\mu_{r\text{eff}}$  of a YBaCuO coplanar line on  $\text{LaAlO}_3$  substrate independence on the temperature (triangles and squares, respectively) and calculation of  $\mu_{r\text{eff}}(T)$  (solid line) with:  $w = 36 \mu\text{m}$ ,  $w/d = 0.18$  and  $\lambda_L(0) = 286 \text{ nm}$

(9), because the magnetic penetration depth shows a strong temperature dependence near  $T_C$ . The solid line represents the result of a simulation for  $\mu_{r\text{eff}}$  with  $\lambda_L(0) = 286 \text{ nm}$  and  $T_C = 90.3 \text{ K}$  using (8) and (9). It is almost identical to the measured values. The determination of the two constants  $\epsilon_{r\text{eff}}$  and  $\mu_{r\text{eff}}$  shows that the temperature dependence of  $Z_L$  and  $\beta$  is only caused by a variation of  $L'$ , because  $C'$  is nearly independent of the temperature.

Additionally, the frequency dependence of the normalized propagation coefficient  $\beta/\beta_0$  has been considered.  $\beta/\beta_0$  was determined from the resonant frequencies of resonators at the different modes  $n$  and showed no dependence on the frequency. Therefore the product  $\mu_{r\text{eff}} \cdot \epsilon_{r\text{eff}}$  does not depend on the frequency, either. Little is known about the frequency dependence of each constant  $\mu_{r\text{eff}}$  and  $\epsilon_{r\text{eff}}$ . The effective permeability  $\mu_{r\text{eff}}$  should be independent of the frequency according to (8). The relative dielectric constant  $\epsilon_r$  of  $\text{LaAlO}_3$  shows no frequency dependence between 5 GHz and 40 GHz [13]. So  $\epsilon_{r\text{eff}}$ , which is only determined by the spatial arrangement of the dielectric [see (7)], might be approximately independent of the frequency in the used frequency range. With these results, it is possible to calculate exact dimensions for given  $Z_L$  and  $\beta$  of coplanar YBaCuO waveguides according to (5) and (6). In the following, two simulations (Figs. 7 and 8) are illustrated with  $\epsilon_{r\text{eff}} = 11.4$ ,  $t = 180 \text{ nm}$  and  $\lambda_L(T) = 400 \text{ nm}$ , which results at 77 K from  $\lambda_L(0) = 286 \text{ nm}$ .

Fig. 7 shows  $w/d$ , which is necessary to obtain a desired  $Z_L$ , in dependence on  $w$ . In order to realize an exactly defined line impedance for arbitrary line widths, the required  $w/d$  must be increased significantly to narrow line widths. With this simulation it is possible to design the exact geometrical dimensions of transmission lines with given line impedances. This is important for technical applications, if an optimal matching of transmission lines is desired.

By changing the line geometry to obtain a desired  $Z_L$ , the propagation coefficient  $\beta$  is also changed. The normalized propagation coefficient  $\beta/\beta_0$ , which results from this redimensioning, is calculated according to (6) and (8) in dependence on the line width  $w$  for different line impedances  $Z_L$  using the same material parameters. The results are presented in Fig. 8, where the normalized propagation coefficient  $\beta/\beta_0$  is

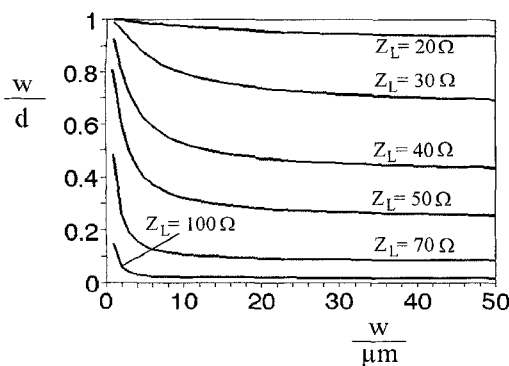


Fig. 7.  $w/d$  ratio necessary to obtain a constant line impedance  $Z_L$  at a given line width  $w$ . Calculated for coplanar waveguides made of YBaCuO with:  $t = 180$  nm,  $\epsilon_{r\text{eff}} = 11.4$  and  $\lambda_L(T) = 400$  nm.

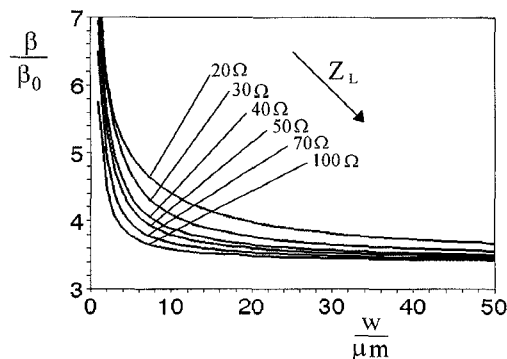


Fig. 8. Normalized propagation coefficient  $\beta/\beta_0$ , which results from redimensioning of the coplanar waveguides to obtain a constant line impedance  $Z_L$ . Calculated in dependence on the line width  $w$  for coplanar waveguides made of YBaCuO with  $t = 180$  nm,  $\epsilon_{r\text{eff}} = 11.4$ , and  $\lambda_L(T) = 400$  nm.

shown. For line widths above  $w = 20$   $\mu\text{m}$ ,  $\beta/\beta_0$  is nearly independent of the line width. Increasing  $w$ , the curve takes an asymptotic course to  $\beta/\beta_0 = \sqrt{\epsilon_{r\text{eff}}} = \sqrt{11.4}$ . Contrary to perfect conductors, the normalized propagation coefficient changes with the line impedance. Using constant line widths,  $\beta/\beta_0$  increases with decreasing line impedances. At narrow lines below  $w = 10$   $\mu\text{m}$ ,  $\beta/\beta_0$  of the redimensioned lines shows a significant increase. The response of  $\beta/\beta_0$  is an important aspect to dimension matched resonators with a special resonant frequency. In order to realize resonators for arbitrary line widths with  $Z_L = \text{constant}$ , the necessary line length  $l$  decreases with decreasing line widths. Just at narrow line widths a strong shortening of the line length is required to achieve the desired resonant frequency.

Both figures show that it is very difficult to realize narrow waveguides ( $w < 4$   $\mu\text{m}$ ) with exact line parameters, because they show a strong dependence on small variations of their geometrical dimensions which are caused by the patterning method.

## V. CONCLUSION

This paper described microwave measurements on narrow coplanar structures made of YBaCuO thin films with line widths from 1 to 40  $\mu\text{m}$ . From these measurements the line impedance and the propagation coefficient were investigated

in dependence on the line width. In spite of a constant ratio between line width and the distance of the ground layers, both quantities increase significantly to narrow line widths. The examinations showed, that it is important to introduce the effective permeability  $\mu_{r\text{eff}}$  in order to give an exact description of the microwave properties of superconducting waveguides. The effective dielectric constant  $\epsilon_{r\text{eff}}$  and the effective permeability  $\mu_{r\text{eff}}$  have been determined from the measurements. The results can be explained by an analytic expression for  $\mu_{r\text{eff}}$ , which uses only the magnetic penetration depth  $\lambda_L$  and the geometric dimensions of the coplanar structures.

While the value of  $\epsilon_{r\text{eff}}$  is independent of the dimensions of the line and the temperature,  $\mu_{r\text{eff}}$  shows a significant increase to narrow line widths and increasing temperatures. With the results of this work it is possible to predict the line impedance  $Z_L$  and the propagation coefficient  $\beta$  of coplanar structures if the YBaCuO parameters like penetration depth  $\lambda_L$  and the thickness  $t$  are known. Therefore, transmission lines and resonators can be designed exactly. This is important for technical applications, where exactly defined properties like line impedance and resonant frequency are needed.

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**Jürgen Berntgen** was born in Köln, Germany, in 1968. After studying electrical engineering at the RWTH Aachen, Germany, he received the Dipl.-Ing. degree in August 1994.

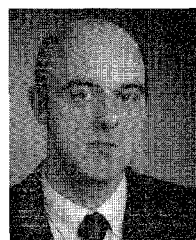
During his work at the Institut für Halbleitertechnik, RWTH Aachen, he examined the microwave properties of narrow waveguides made of YBaCuO. He is now with the Institut für Halbleitertechnik, and concentrates on the research of passive microwave devices made of YBaCuO.



**Eberhard Waffenschmidt** was born in Duisburg, Germany, on April 17, 1964. After studying electrical engineering at the RWTH Aachen, Germany, he received the diploma degree in May, 1991. He concentrated his work on passive microwave devices made of YBaCuO and received the Ph.D. degree in October, 1995.

During his work he examined vapor pressures of precursors used for metalorganic chemical vapor deposition (MOCVD) of high temperature superconductors (HTSC). He is now with the Institut für

Halbleitertechnik, RWTH Aachen, Germany. Together with two colleagues he built up the HTSC group evaluating the use of superconducting YBaCuO thin films made by MOCVD for microelectronic applications.



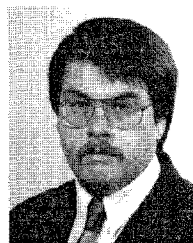
**Jürgen Musolf** was born in Krefeld, Germany, in 1963 and studied physics at the RWTH Aachen, Germany, where he received his diploma degree in 1990. The topic of his diploma thesis was the metalorganic molecular beam epitaxy (MOMBE) of GaAs and AlGaAs. He received the Ph.D. degree at the Institut für Halbleitertechnik, RWTH Aachen, in October 1994.

His current research is on the metalorganic chemical vapor deposition (MOCVD) growth of YBaCuO. He is now working for Superconductor Technologies, USA where he is developing the MOCVD-technique to become the production method of choice for the growth of high quality HTSC.



**Xiaoyin He** (M'93) received the B.Sc. and M.Sc. degrees in electronic engineering from Tsinghua University, Beijing in 1983 and 1986, and the Ph.D. degree in electrical engineering from Aachen Technical University (RWTH Aachen), Aachen, Germany, in 1995.

From 1986 to 1989 he was with Institute of Microelectronics, Tsinghua University, working on the silicon oxide thin film and EEPROM device and circuit. From 1990 to 1995 he was with IHTI, RWTH Aachen, working on the research and development of YBaCuO thin films and superconducting bolometers. In 1995 he joined superconducting core technologies.



**Michael Heuken** was born in Oberhausen, Germany on November 17, 1961. He received the Diplom-Ingenieur degree and the Dr.-Ing. degree in electrical engineering from Duisburg University in 1985 and 1989, respectively. In 1994, he finished his Habilitation degree in semiconductor technology and devices with a thesis on ZnSSe-based optoelectronic devices.

He joined the Institut für Halbleitertechnik at RWTH Aachen as Senior Engineer. He is working in the field of metalorganic vapor phase epitaxy for electronic and optoelectronic devices. Now he teaches semiconductor technology and devices as well as circuits for communication systems.

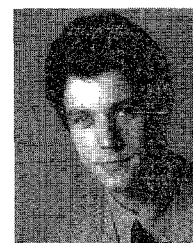
Dr. Heuken is member of DGKK and VDE.



**Klaus Heime** (SM'78) was born in Heidelberg, Germany, on August 1, 1935. He received the Diplom-Physiker degree in 1961 from the University of Bonn, and the D. Science degree from the Technische Hochschule, Darmstadt, Germany, in 1967.

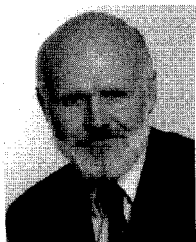
From 1961 to 1971 he was a Member of the Research Institute of the Telecommunications Central Office (Federal Post Office) in Darmstadt, working in the field of semiconductor technology (thin-film devices, GaAs technology, Schottky diodes, and Gunn devices and circuits). In 1971 he became Professor at the Technische Hochschule, Aachen, Germany, and in 1975 he joined the Universität-Gesamthochschule-Duisburg, Germany. Since 1989 he has been Director of the Institute of Semiconductor Technology, Technical University of Aachen. He teaches solid-state electronics while his research interest are in the field of III-V semiconductor technology and integrated circuits.

Dr. Heime is a member of the German Physical Society, Informationstechnische Gesellschaft (ITG) im VDE, where he heads a chapter semiconductor devices, Gesellschaft für Mikroelektronik (VDI/VDE); The American Physics Society.



**Stefan Hofschien** was born in Sevelen, Germany in 1967. He studied electrical engineering at the Technical University of Aachen, Germany with the main focus on solid state electronics and received his diploma in December 1991.

From May 1991 to March 1992 he was employed as a working student by the Daimler-Benz Research Center in Ulm, Germany and engaged in the design and microwave characterization of passive, superconducting MMIC devices. Since April 1992 he has been with the faculty of electrical engineering at the Gerhard Mercator University Duisburg, Germany as a Research Assistant and is currently working toward the Ph.D. degree. His research interests are the development of simulation methods and the design of passive, lossy, super- and normalconducting MMIC components.



**Ingo Wolff** (M'75-SM'85-F'88) was born in Köslin, Germany, in 1938. He studied electrical engineering at the Technical University of Aachen and received the Dipl.-Ing. degree in 1964. In 1967 he received the Doctoral degree and in 1970 the habilitation degree, also from the Technical University of Aachen, Germany.

From 1970 to 1974 he was a Lecturer and Associate Professor for high-frequency techniques in Aachen. Since 1974 he has been a Full Professor of electromagnetic field theory at the University of Duisburg, Duisburg, West Germany. His main areas of research are electromagnetic field theory applied to the computer-aided design of MIC's and MMIC's, millimeter-wave components and circuits, and the field theory of anisotropic materials. Since 1992 he has lead the Institute of Mobile and Satellite Communication Techniques, Kamp-Lintfort, in parallel to his university position. This institute intensively works in the area of mobile communication techniques, microwave and millimeter wave techniques, antenna techniques, and electromagnetic compatibility techniques.